AD-A104 534 CALIFORNIA UNIV BERKELEY DEPT OF PHYSICS F/G 9/5
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1981 W R MCGRATH, P L RICHARDS, A D SMITH N00014-75-C-0496

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Submitted to Applied Physics Letters

Contract N00014-75-C-0496

LEVEL II

AD A104534

LARGE GAIN, NEGATIVE RESISTANCE, AND OSCILLATIONS IN SUPERCONDUCTING QUASIPARTICLE HETERODYNE MIXERS

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ABSTRACT

We have measured the properties of a superconductor-insulator-superconductor (SIS) quasiparticle mixer which is operated in the quantum limit. Single sideband conversion gain larger than +4 dB was observed at 36 GHz with a mixer noise temperature T=9±6K, which is to be compared with the (Planck) quantum limit $\hbar\omega/(k\ln 2) \approx 2.5$K. Complete 3-port mixer calculations are presented which are in good agreement with the gain measurements. Negative resistance was observed which implies that arbitrarily large gain is available.

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Low noise microwave heterodyne receivers which use a superconductor-insulator-superconductor (SIS) tunnel junction in the quasiparticle regime for the mixing element are being actively investigated. Mixing has been reported over the frequency range from 10 to 115 GHz in lead-alloy tunnel junctions[1-7]. Arrays of junctions[3,7] in series have been successfully used in addition to single junctions. The conversion efficiencies measured at 36 and 74 GHz are larger than the limit obtainable from resistive mixers[4,7]. The mixer noise temperature measured at 36 GHz is comparable with the quantum limit $T_M \approx \hbar\omega/k\ln 2$.

The photon assisted tunneling theory of quasiparticle microwave mixers has been explored extensively[8,9]. Calculations predict[9,10] that there is a wide range of experimental parameters for which SIS mixers will have large conversion gain and noise comparable to the quantum limit[10]. The importance of conversion gain is evident from the equation for the noise temperature T_R of a heterodyne receiver with a coupled antenna noise T_A,

$$T_R = T_A + T_M + T_{IF}/G, \tag{1}$$

where the noise temperature T_{IF} of the intermediate frequency amplifier is reduced by the mixer conversion gain G. We describe here the first observation of coupled IF output power from a quasiparticle mixer which is larger than the available input signal. The values of RF source impedance and LO power at which this gain is observed are in good agreement with the results of 3-port mixer calculations.

In calculations of SIS mixer performance[9,10] large gain is accompanied by large values and even negative values of dynamic resistance R_D at the

IF frequency. Although negative resistance due to Josephson pair tunneling is commonplace, RF-induced negative quasiparticle resistance has not been previously reported. We have observed arbitrarily large (and negative) values of R_D which implies that arbitrarily large gain is available.

For our experiments we have made tin-tin oxide-tin SIS junctions on Si substrates using photoresist bridge masks[12]. Typical junctions have areas of $\sim 10\ \mu m^2$ and critical current densities of $\sim 250\ Acm^{-2}$. The tunneling I-V curves, shown in Fig. 1, are very sharp. Using these junctions we are able to observe strong quantum effects at an operating frequency low enough that accurate microwave measurements are relatively easy.

The apparatus used for mixer measurements is the same as that described in previous publications[1,4]. The helium bath temperature was generally 1.5K, although mixer efficiency varied only weakly with temperature for $T \lesssim T_c/2$. The Sn junction was placed in the E-field direction across a Ka-band waveguide. Radio frequency matching was obtained by adjusting a screw tuner and a sliding backshort. Room temperature radiation was reduced by a cold 18 dB attenuator in front of the mixer. The signal and LO power were obtained from carefully calibrated 36 GHz oscillators. Conversion efficiency measurements were made with an IF frequency of 50 MHz and noise measurements with an IF band from 15 to 130 MHz.

In Fig. 1(a) we show a series of measured I-V curves for a 22Ω junction for values of RF source impedance in the region of optimum mixer operation. A slight increase in RF source impedance produced I-V curves with both larger values of dynamic resistance and also negative resistance on the photon assisted tunneling steps as is shown in Fig. 1(b). The region to the left of the dashed line in each part of the figure is very sensitive to an applied magnetic field large enough to produce one flux quantum in the junction. The complicated structure seen there, including regions of negative resistance, arises primarily from pair tunneling effects. The region to the right of the dashed line, on the other hand, is essentially independent of field. We believe that the negative resistance in this latter region arises from the quasiparticle current. In a separate publication we show that this negative resistance can be obtained from a simple analytic model for a pumped junction in the quantum limit with finite RF source resistance[13]. A magnified view of a negative resistance region measured with a 130Ω DC bias resistance is shown in the inset of Fig. 1(b). Oscillations were observed at harmonics of ∿5 MHz when the junction was biased in a negative resistance region.

Substantial mixer gain was observed by using I-V curves similar to those shown in Fig. 1(a). The best experimental values are +4.3±1 dB for the upper sideband and +3.4±1 dB for the lower sideband. The experimental curves from which these results were derived are shown in Fig. 2. The values of conversion efficiency were obtained by dividing the output power coupled to the IF amplifier by the monochromatic signal power (which was <1.5 pW) entering the cryostat. Corrections were then made for the measured attenuation of the waveguide (-1.9 dB), the cold attenuator (-17.8 dB) and the IF cable (-0.1 dB). No corrections were made for power lost due to RF or IF mismatch.

To model mixer operation, we propose the equivalent IF output circuit shown as an inset in Fig. 3. The mixing action produces a current amplitude I_{IF}, which is shunted by the dynamic resistance of the junction R_D and the IF amplifier impedance R_L. For a range of parameters around the maximum gain point, I_{IF} is relatively independent of R_D and V_{DC}. Theoretical support for this model has been published previously[5,11]. For matched output ($R_D = R_L$), the available output power is $I_{IF}^2 R_D/8$. In our experiment the output was not matched, so the coupled power is the product of the available power with the mismatch factor $4R_D R_L/(R_D+R_L)^2$. In order to test this model, the coupled IF output power was measured by varying V_{DC} along a single step and was plotted as crosses in Fig. 3 as a function of the measured R_D. The coupled gain numbers were then divided by the mismatch factor to obtain values of the available power shown as circles in Fig. 3. Good agreement is obtained with our assumption that the available output power is proportional to R_D. Since the dynamic resistance can be infinite, this model predicts that gain is limited only by the load resistance R_L. An improvement of ~1 dB in measured gain was obtained by using a capacitively loaded length of transmission line to increase R_L above the 50Ω input resistance of the IF amplifier.

To determine the mixer noise temperature, the noise contribution of the IF amplifier and of the lossy coaxial IF cable were measured in separate experiments. Subtracting these IF noise sources left the mixer output noise, which was converted into an equivalent input noise by dividing by the SSB gain for the upper sideband. Finally, blackbody radiation T_A present in the waveguide was subtracted leaving the intrinsic mixer input noise temperature of 9±6K. The noise was measured under the operating conditions for maximum gain.

Mixer performance depends critically on the RF source embedding admittance Y_S. The experimental value of Y_S can be determined from a pumped I-V curve[4,10]. For our junctions the estimated relaxation parameter $\omega R_N C \approx 7$ at 36 GHz. Therefore, the junction capacitance provides a very low impedance termination at harmonic frequencies, and we can use the 3-port Y-mixer model. Estimates for Y_S were obtained both by the published method[14] and also by computing a pumped I-V curve as a function of source admittance and obtaining the source parameters from a least-squares fit to the corresponding experimental curve. The resulting source admittance is $Y_S = (0.07\pm0.01) + j(0.01\pm0.02)$ mho. In this fitting the geometrical capacitance of the junction, which contributes $Y=j0.3$ mho, is regarded as part of the source. Since the deduced value of Y_S is nearly real, it is clear that the matching introduced reactive elements which resonated out much of this junction capacitance at the signal frequency. The linear portion of the quantum reactance of the junction corresponds to $Y\sim0.01$ mho, so is not quantitatively important.

Mismatch at the IF amplifier also strongly influenced our measured values of gain. For the purposes of these comparisons we define gain in terms of the IF power coupled into a 50Ω load. In Fig. 4 we show calculated gain contours for a bias voltage $V = (2\Delta-\hbar\omega/2)/e$ on the first photon assisted tunneling step below $2\Delta/e$. The contours are plotted as a function of normalized RF source conductance G_S and normalized P_{LO} so as to show how the gain depends upon these important experimental parameters. It is assumed that all linear RF reactances have been tuned out. Qualitative differences between Fig. 4 and previously published gain contours[10] are due to our use of a fixed value of R_L. In addition to the gain contours in Fig. 4 we show the values of the independent variables for which large gain was actually observed. The agreement with theory is quite good. However, there is a significant discrepancy in the amount of gain which is predicted. It is to be expected that the 3-port model overestimates the gain, and that inclusion of dissipation at the harmonic frequencies would improve the agreement between experiment and theory[4].

When one of our junctions is operated as a direct detector, it is a microwave photodiode with current responsivity $e/\hbar\omega$ [8,15]. The direct detection response can thus be used as a check on the calibration of the RF sources used for mixer tests. The calibration obtained in this way corresponds to less gain and less noise than the direct calibration. This uncertainty is included in the error limits given for the values of gain and noise reported above. We have not accepted the direct detection responsivity as the primary calibration because the possibility exists of gain associated with quasiparticle injection[16]. Because of the small leakage currents in our junctions, the shot noise in the direct detector is small. The noise-equivalent-power of the square-law detector can be predicted to be $NEP=1\times10^{-16}$ W $Hz^{-1/2}$ for a rapidly modulated signal.

In conclusion, we have observed that SIS heterodyne mixers operated in the quantum limit show large gain with low noise, as well as negative resistance. This behavior is in good agreement with the predictions of quantum mixer theory. We have also improved the performance of the SIS direct detector.

We are grateful for helpful discussions with T. M. Shen, C. C. Chi and K. Chen. One of us (H.v.K.) thanks the Netherlands Organization for the Advancement of Pure Research (ZWO) for a grant. A portion of this research was supported by the U.S. Office of Naval Research.

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FIGURE CAPTIONS

Fig. 1 Measured I-V curves for our Sn junction at 1.5K (a) without and (b) with negative resistance due to quasiparticle tunneling. The higher curves are for 36 GHz P_{LO} increasing upward in 2 dB steps starting with $\sim$0.2 nW. To the left of the dashed line the curves are very sensitive to magnetic field, and so contain important contributions from pair tunneling. To the right of the dashed line the curves are essentially field independent, suggesting that only quasiparticle effects are important. The inset in (b) shows a region of negative quasiparticle resistance in detail.

Fig. 2 Performance curves for an SIS mixer measured for settings of the screw tuner and backshort which maximized conversion gain. The signal and noise curves were measured at the output of the IF amplifier. Both of these curves are strongly influenced by the bias voltage dependence of the IF mismatch factor. There is a minimum in the noise whenever R_D is 50Ω.

Fig. 3 The equivalent output circuit of an SIS mixer is shown in the inset. The mixer is modeled as a current generator I_{IF} shunted by the dynamic resistance R_D of the pumped junction. The IF amplifier is represented by the load resistance R_L. The conversion gain of a fundamental mixer measured for various values of DC bias on the first photon peak below $2\Delta/e$ is shown as a function of the measured R_D. The coupled G_{SSB} values (crosses) saturate for large values of R_D because of IF impedance mismatch. When corrected for mismatch (circles), the gain is nearly proportional to R_D, in agreement with the equivalent circuit for constant I_{IF}.

Fig. 4 Gain predicted from a 3-port quantum Y-mixer model using the experimental DC I-V curve. The junction is biased on the first photon assisted tunneling step below $2\Delta/e$. Gain contours are shown as a function of the normalized source conductance and the normalized available LO power. The range of parameters for which large gain was observed are given by the box.

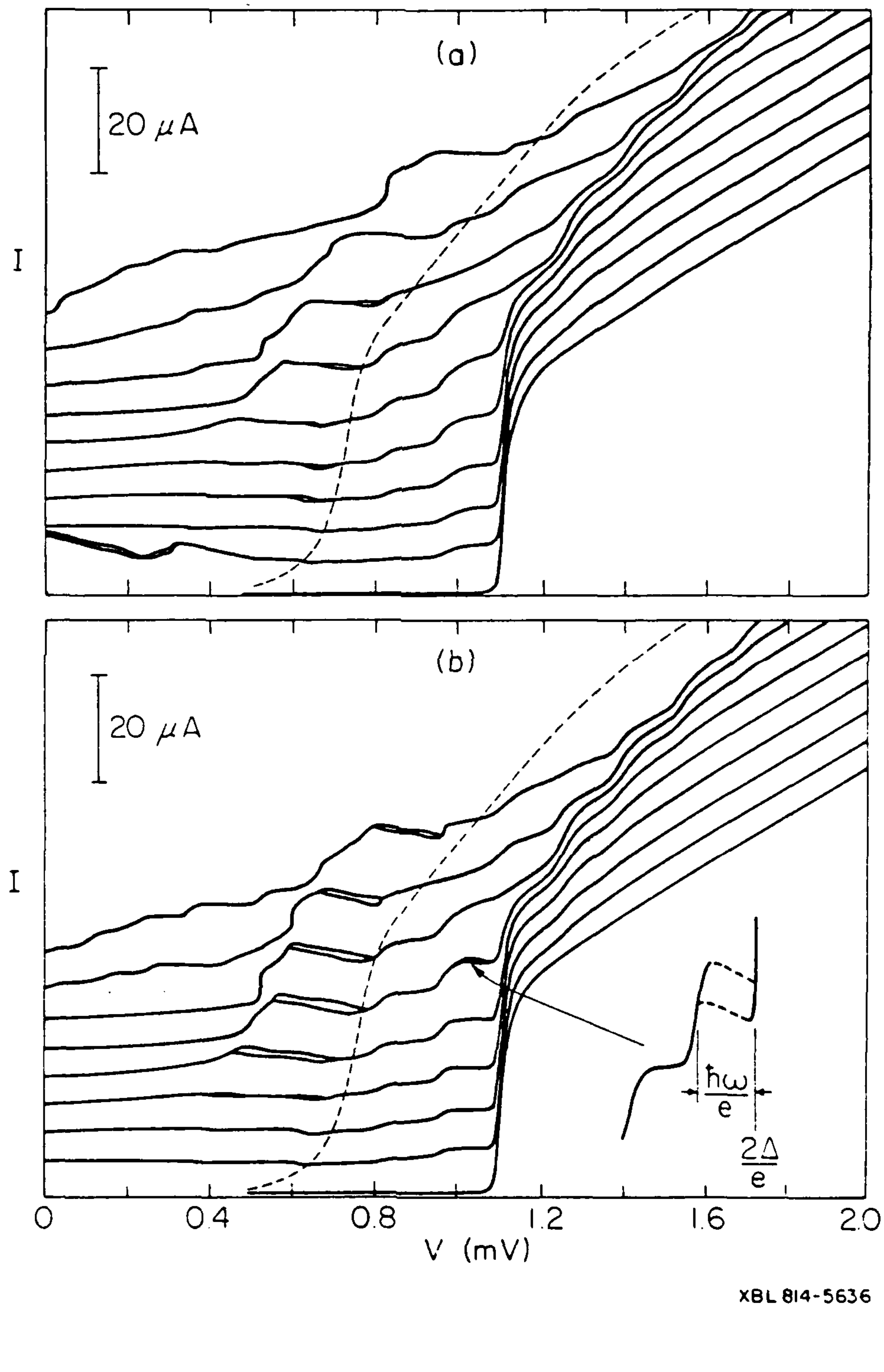

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Fig. 1

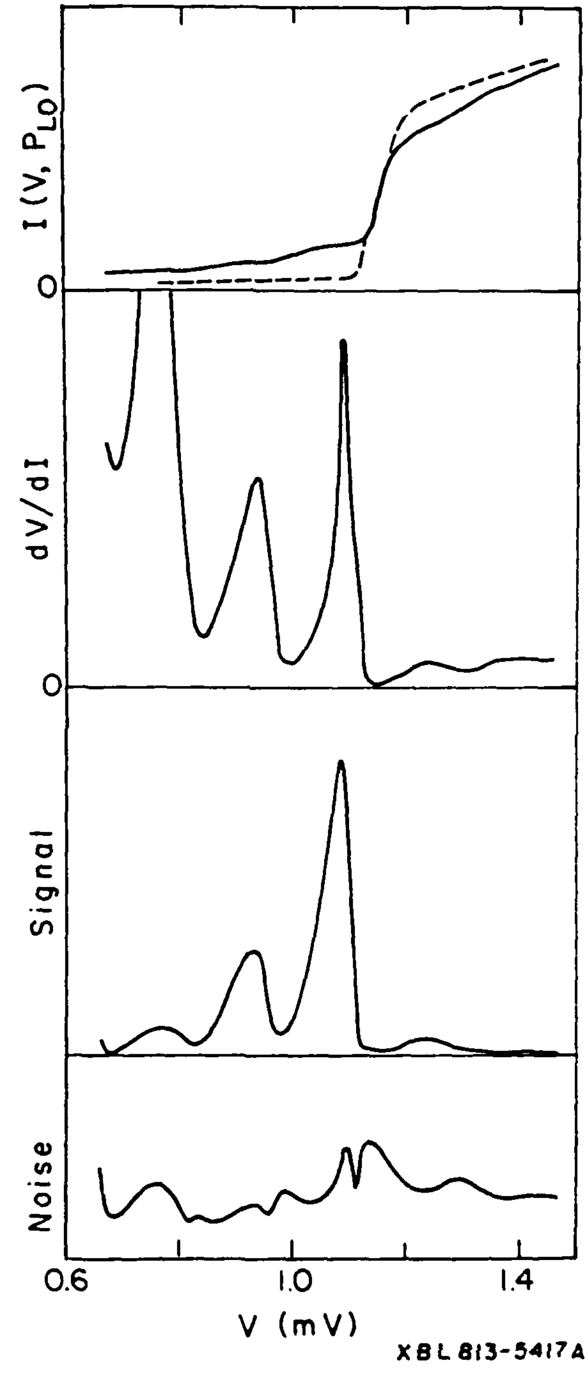
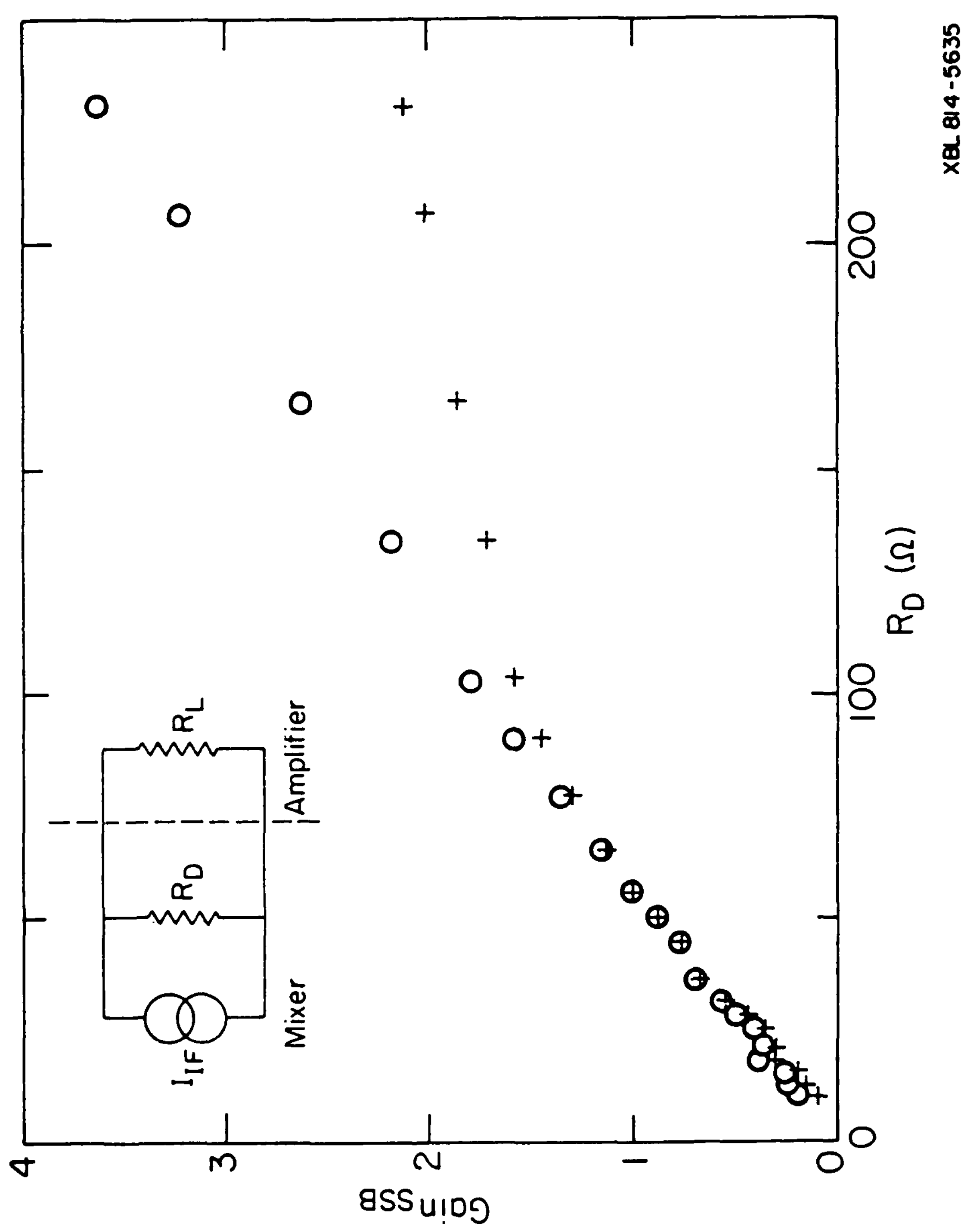


Fig. 2



Gain SSB
R_D (Ω)
I_{IF}
R_D
R_L
Mixer
Amplifier
0
1
2
3
4
0
100
200
XBL 814-5635

Fig. 3

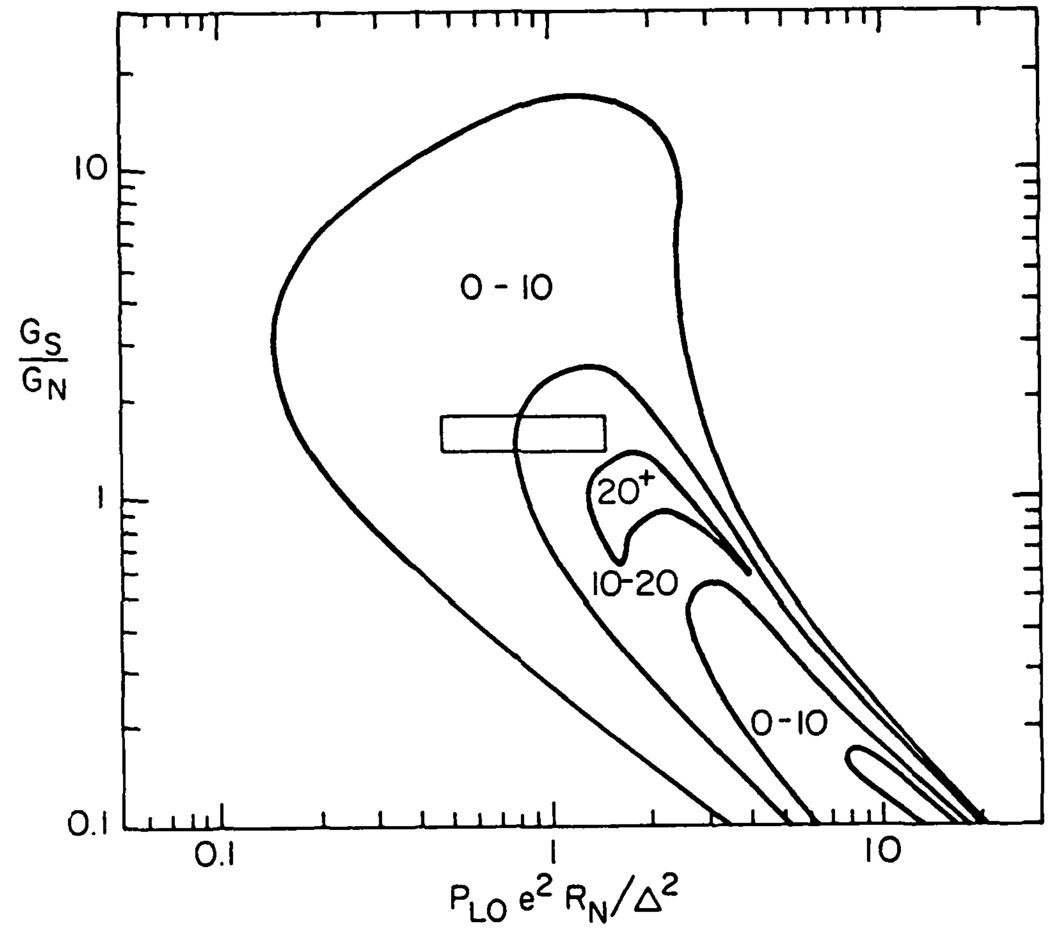


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Fig. 4